TOSHIBA Bi-CMOS Integrated Circuit Silicon Monolithic

# **TB6586FG**, **TB6586AFG**

Three-Phase Full-Wave Brushless Motor Controller

#### Features

- Upper-phase PWM control
- Built-in triangular-wave generator
- Support of a bootstrap circuit
- Built-in Hall amplifier (support of a Hall element)
- Selectable 120°/150° energization
- Built-in lead angle control function
- Overcurrent protection signal input pin (VRS = 0.5 V (typ.))
- Built-in regulator (V<sub>refout</sub> = 5 V (typ.), 35 mA (max))
- Operating supply voltage range:  $V_{CC}$  = 6.5 to 16.5 V,  $V_M$  = 4.5 to 16.5 V
- The TB6586FG and TB6586AFG differ in the number of pulses per revolution: TB6586FG: 1 pulse / electrical angle: 360° TB6586AFG: 3 pulses / electrical angle: 360°



Weight: 0.27 g (typ.)

About solderability, following conditions were confirmed

• Solderability

- (1) Use of Sn-37Pb solder Bath
  - solder bath temperature = 230°C
  - dipping time = 5 seconds
  - the number of times = once
  - use of R-type flux
- (2) Use of Sn-3.0Ag-0.5Cu solder Bath
  - · solder bath temperature =  $245^{\circ}C$
  - dipping time = 5 seconds
  - the number of times = once
  - use of R-type flux

## <u>TOSHIBA</u>

### **Pin Description**

Pin No.	Symbol	Description
1	V <sub>SP</sub>	Speed control
2	HUP	U-phase Hall signal input (+) pin
3	HUM	U-phase Hall signal input (–) pin
4	HVP	V-phase Hall signal input (+) pin
5	HVM	V-phase Hall signal input (–) pin
6	HWP	W-phase Hall signal input (+) pin
7	HWM	W-phase Hall signal input (-) pin
8	V <sub>refout</sub>	Outputs reference voltage signal (5 V / 35 mA)
9	LA	Lead angle setting signal input pin (30° / 4 bits)
10	GND	Ground pin
11	CW/CCW	Rotation direction signal input pin
12	OSC/C	Connect to condenser for PWM oscillator
13	OSC/R	Connect to resistor for PWM oscillator
14	RS	Overcurrent protection (0.5 V)
15	RESET	Energization width toggle pin (Low: 150°, High; Reset, 6.35 V: 120°)
16	V <sub>CC</sub>	Power supply
17	VM	Input for output power
18	UL	U-phase output pin (Low)
19	VL	V-phase output pin (Low)
20	WL	W-phase output pin (Low)
21	UH	U-phase output pin (Low)
22	VH	V-phase output pin (Low)
23	WH	W-phase output pin (Low)
24	FG	Output of number of pulses per revolution (FG: 1 pulse / electrical angle; AFG: 3 pulses / electrical angle)

### Pin Layout



## Input/Output Equivalent Circuits

Pin Description	Symbol	Input/Output Signal	Input/Output Internal Circuit
Positional signal input pin	HUP HUM HVP HVM HWP HWM	Analog/Digital Hysteresis ± 7.5 mV (typ.)	Vrefout Vrefout
Speed control signal input pin	V <sub>SP</sub>	Analog Input range 0 to 7 V	
Rotation direction signal input pin L: Forward (CW) H: Reverse (CCW)	cw/ccw	Digital L: 0.8 V (max) H: V <sub>refout</sub> – 1 V (min) (Test input If CW/CCW = 6.35 V (typ.) or higher, the system resets Hysteresis 150 mV (typ.)	V <sub>CC</sub> 65 kΩ CW/CCW Gy S0 CW/CCW CW/CCW Cy S0 CW/CCW
Reset input L: 150° turn-on mode H: Reset	RESET	Digital L: 0.8 V (max) H: V <sub>refout</sub> – 1 V (min) If RESET = 6.35 V (typ.) or higher, then 120° energization drive is selected Hysteresis 150 mV (typ.) During a reset: Output OFF (all phases Low). The internal counter continues to operate.	$V_{CC}$ $65 k\Omega$ $G_{W}$ $G_{V}$
Lead angle setting signal input	LA	Analog Input range 0 to 5.0 V (V <sub>refout</sub> ) Electrical angle 0°~28° can be divided into 16 by 4-bit data. Lead angle 0°: LA = 0 V (GND) Lead angle 28°: LA = 5 V (V <sub>refout</sub> )	Vrefout 100 kΩ 0 0 0 0 0 0 0 0 0 0 0 0 0

Pin Description	Symbol	Input/Output Signal	Input/Output Internal Circuit			
Overcurrent protection signal input	RS	Analog Analog filter 0.5 μs (typ.) If RS = 0.5 V (typ.) or higher, UH, VH and WH pin goes low (released at carrier cycle)	Vrefout Vrefout Vrefout C C C C C C C C C C C C C			
Reference voltage signal output pin	Vrefout	5.0 ± 0.5 V (35 mA) 5.0 ± 0.3 V (15 mA)	Vcc VccVcc			
Rotational frequency output	FG	Digital Push-pull output (± 2 mA (max)) TB6586FG: 1 pulse / electrical angle of 360° TB6586AFG: 3 pulses / electrical angle of 360°	Vrefout Vrefout Vrefout 100 Ω			
Energization signal input	UH UL VH VL WH WL	Push-pull output (± 3 mA (max))	↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓			

### **Block Diagram**



### **Absolute Maximum Ratings**

Characteristic	Symbol	Rating	Unit	
	V <sub>CC</sub>	18	V	
Supply Voltage	VM	18	v	
	V <sub>IN1</sub> -0.3 to 8 (Note 1)			
Input voltage	V <sub>IN2</sub>	-0.3 to 8.5 (Note 2)	V	
	LA	-0.3 to V <sub>refout</sub> + 0.3		
Turn-on signal output current	IOUT	3	mA	
Supply voltage	Pa	0.8 (Note2)	W	
Supply vollage	FD	1.0 (Note 3)		
Operating temperature	T <sub>opr</sub> –30 to 85		ŝ	
Storage temperature	T <sub>stg</sub>	-55 to 150	C	

Note 1: CW/CCW, RESET

Note 2: VSP

Note 3: No heatsink

Note 4: When mounted on a PCB ( $50 \times 50 \times 1.6$  mm, Cu 10%)

#### **Operation Conditions (Ta = 25°C)**

Characteristic	Symbol	Min	Тур.	Max	Unit
Supply voltage	V <sub>CC</sub>	6.5	15	16.5	V
Supply voltage	VM	4.5	_	16.5	V
Oscillation frequency	Fosc	2	5	8	MHz

## Electrical Characteristics (Unless otherwise specified Ta = $25^{\circ}$ C, V<sub>CC</sub> = 15 V, V<sub>M</sub> = 5 V)

Characteristic		Symbol		Test Circuit	Test Condition	Min	Тур.	Max	Unit
Supply current		ICC			$V_{refout} = OPEN, \\ OSC/C = 560 \text{ pF}, \\ OSC/R = 6.2 \text{ k}\Omega$	_	5.5	10	mA
Output current		I <sub>M</sub>		_	Drive output (UH, UL, VH, VL, WH, WL) = OPEN		0.5	1	mA
		I <sub>IN (LA)</sub>			V <sub>IN</sub> = 5 V LA	_	25	50	
		I <sub>IN</sub>	(SP)		V <sub>IN</sub> = 5 V V <sub>SP</sub>	_	35	70	
Input current		IIN (RESET) IIN (CW) IIN (RS)			V <sub>IN</sub> = 5 V RESET	_	25	50	μΑ
					V <sub>IN</sub> = 5 V CW/CCW	_	25	50	-
					V <sub>IN</sub> = 0 V RS	_	-25	-50	
		V <sub>IN1</sub>	High	_	CW/CCW, RESET	V <sub>refout</sub> – 1		V <sub>refout</sub>	V
			Low			0	_	0.8	
		VI	N2	—	RESET: 120° turn-on mode	6.0	6.35	6.7	
Input voltage		V <sub>RST1</sub>		—	CW/CCW: System reset	6.0	6.35	6.7	v
pat tonago		V <sub>RST2</sub>		—	RESET: Power off reset	2.2	—	V <sub>refout</sub>	
			Н		PWM ON duty 95%	5.1	5.4	5.7	
		V <sub>SP</sub>	м	] —	Refresh $\rightarrow$ Start motor operation	1.8	2.1	2.4	V
			L		Energization OFF $\rightarrow$ Refresh	0.7	1.0	1.3	
	Input sensitivity	V <sub>S</sub> V <sub>W</sub> VH (1)			Differential input	40		—	mVpp
Hall element input	Common mode					1.5	_	3.5	V
	Input hysteresis				(Note)	±4.5	±7.5	±10.5	mV
Input hysteresis voltage		VH <sub>(2)</sub> VH <sub>(3)</sub>			RESET: Reset $\leftrightarrow$ 120°energization(Note)		0.15		- V
					$\begin{array}{c} CW/CCW: CCW \leftrightarrow Reset \\ (Note) \end{array}$		0.15		
Input delay		Tf	RS		$RS \to Output \; OFF$	_	2.2		μs
		VOUT (15) – H VOUT (15) – L VOUT (5) – H VOUT (5) – L VFG (H)		- - 	$I_{OUT} = 3 \text{ mA}, V_M = 15 \text{ V}$	13	14.2	—	- V
					$I_{OUT} = 3 \text{ mA}, V_M = 15 \text{ V}$	_	0.8	1.2	
					$I_{OUT} = 2 \text{ mA}, V_M = 5 \text{ V}$	4.0	4.2		
					$I_{OUT} = 2 \text{ mA}, V_M = 5 \text{ V}$		0.8	1.0	
Calput Voltage					I <sub>OUT</sub> = 2 mA FG	4	_		
		VFC	G (L)	-	I <sub>OUT</sub> = 2 mA FG	_	_	1.0	-
		Vref	iout1		I <sub>OUT</sub> = 15 mA V <sub>refout</sub>	4.7	5.0	5.3	
		Vref	out2		I <sub>OUT</sub> = 35 mA V <sub>refout</sub>	4.5	5.0	5.5	
Output leakage cur	rent	١L	(H)		$V_{OUT} = 0 V$	_	0	1	
Output leakage current		١L	(L)		V <sub>OUT</sub> = 15 V	_	0	1	μΑ
Electrical current detector		Vi	RS		RS	0.46	0.5	0.54	V
Lead angle correction		T <sub>LA (0)</sub>		_	LA = 0 V or open, Hall IN = 100 Hz	_	0	_	
		T <sub>LA (2.5)</sub>			LA = 2.5 V, Hall IN = 100 Hz		17		0
		T <sub>LA (5)</sub>			LA = 5 V, Hall IN = 100 Hz	_	28		
V <sub>CC</sub> monitor		V <sub>CC</sub> (H)			Output start operation point	5.7	6.0	6.3	
		V <sub>CC</sub> (L)			No output operation point	4.7	5.0	5.3	V
		VH (4)		—	Input hysteresis width (Note)	—	1.0	—	1

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## TB6586FG/AFG

Characteristic	Symbol	Test Circuit	Test Condition	Min	Тур.	Max	Unit	
PWM oscillator frequency	F <sub>C (20)</sub>	_	$\begin{array}{l} \text{OSC/C} = 560 \text{ pF},\\ \text{OSC/R} = 6.2  \text{k}\Omega \end{array}$	18	20	22	kH7	
(carrier frequency)	F <sub>C (18)</sub>	_	$\begin{array}{l} \text{OSC/C} = 470 \text{ pF},\\ \text{OSC/R} = 8.2  \text{k}\Omega \end{array}$	16.2	18	19.8	KI IZ	
Output duty (max)	T <sub>on</sub> (max)	_	$\begin{array}{l} \text{OSC/C} = 560 \text{ pF}, \\ \text{OSC/R} = 6.2  \text{k} \Omega,  \text{V}_{\text{SP}} = 5.7 \text{ V} \end{array}$	92	95	98	%	

Note: Pre-shipment testing is not performed.

### **Functional Description**

#### 1. Basic operation

At startup, the motor runs at  $120^{\circ}$  energization. When the position detection signal reaches a revolution count of fs = 5 Hz or higher, the rotor position is extrapolated from the position detection signal and output is activated using the lead angle based on the LA signal.

Startup - 5 Hz: 120° energization $fs = f_{0sc} / (120 \times 2^5 \times 2^8)$ 5 Hz or higher: 120° energization or 150° energization \*Approximately 5 Hz if  $f_{0sc} = 5$  MHz.

\*: At 5 Hz or higher, operation is performed in accordance with commands from RESET and LA pins. When the motor is running at 5 Hz or lower and in reverse (in accordance with the timing chart), it will be driven at 120° energization for a lead angle of 0°.

### 2. $V_{SP}$ voltage command signal function

- (1) When voltage instruction is input at  $V_{SP} \leq 1.0$  V: Output is turned off (gate block protection).
- (2) When voltage instruction is input at 1.0 V < V<sub>SP</sub>  $\leq$  2.1 V (refresh operation): The lower transistor is turned on at a regular (carrier) cycle. (ON duty: T<sub>on</sub> = 18/f<sub>osc</sub>)
- (3) When a voltage instruction is input at V<sub>SP</sub> > 2.1 V: The drive signal is output using the energization method configured using the RESET pin.
  - Note: At startup, to charge the upper transistor gate power supply, turn the lower transistor on for a fixed time with 1.0 V <  $V_{SP} \leq 2.1$  V.



\*: The maximum ON duty is  $T_{on} = 95\%$  (typ.) when  $V_{SP} = 5.4$  V (typ.). Example: If  $f_{OSC} = 5$  MHz, then ON time = 48  $\mu$ s (typ.) ( $f_c = 19.8$  kHz) If  $f_{OSC} = 4$  MHz, then ON time = 60  $\mu$ s (typ.) ( $f_c = 15.9$  kHz)

### 3. Function to stabilize the bootstrap voltage

The product is equipped with a bootstrap capacitor charging function that supports the output level of the bootstrap method.

(1) If the VSP input current is  $1.0 \text{ V} < \text{VSP} \leq 2.1 \text{ V}$ , the ON signal is output to the lower phase (UL, VL, WH) based on the carrier cycle. If the output waveform is upper phase (UH, VH, WH), the OFF signal (Low) is output.



Example:  $f_{OSC}=5$  MHz  $T_{On}=3.6~\mu s$ 

(2) If the VSP input current is 2.1 V < VSP and the Hall signal is 5 Hz or less, the upper phase (UH, VH, WH) will perform 120° energization at a PWM that complies with the VSP; and the lower phase (UL, VL, WL) will operate at 120° energization, performing refresh operation based on the OFF timing. (The same drive is executed during "headwind" operation as well.)</p>



 $T_{SP}$ : Variable depending on the V<sub>SP</sub> (the figure above being applicable when V<sub>SP</sub> = 5.4 V (typ.));  $T_{on} = 18/f_{osc}$ ; Td = 18/f<sub>osc</sub>

\*: The lead angle correction (LA pin) function does not operate when the Hall signal is 5 Hz or less. The lead angle correction function also does not operate when in a reverse detection state.

#### 4. Correcting the lead angle

The lead angle can be corrected in the turn-on signal range from 0 to 28° in relation to the induced voltage. Analog input from the LA pin (0 V to 4.3 V divided by 16):

 $0 \,\, V = 0^\circ$ 

4.3 V or higher = 28°

Sample Evaluation Results

Lead Steps LA (V) Angle (° 1 0.00 0.00 0.05 1.93 2 3 0.28 3.79 4 0.59 5.65 5 0.89 7.54 6 1.21 9.43 1.52 11.29 7 8 1.83 13.15 9 2.14 15.08 10 2.45 16.87 11 2.75 18.73 3.06 20.66 12 13 3.37 22.55 14 3.68 24.37 15 3.99 26.16 16 4.30 28.09



#### 5. Setting the carrier frequency

This function involves setting the triangular wave cycle (carrier cycle) necessary for generating PWM signals.

 $Carrier \ frequency: \ f_c = f_{osc}/252 \ (Hz) \qquad \qquad f_{osc} = reference \ clock \ (crystal \ oscillation)$ 

Example: If  $f_{OSC} = 5$  MHz, then  $f_C = 19.8$  kHz If  $f_{OSC} = 4$  MHz, then  $f_C = 15.9$  kHz

#### 6. Position detection pin

The common-mode voltage range is  $V_W = 1.5$  to 3.5 V. The input hysteresis is  $V_H = 7.5$  mV (typ.).



#### 7. Revolution pulse output pin (the difference between TB6586FG and TB6586AFG)

This pin outputs the revolution pulses based on the Hall signal. The TB6586FG outputs one (1) pulse / electrical angle and the TB6586AFG outputs three (3) pulses / electrical angle. In the case of the TB6586FG, this pulse is generated via the U phase Hall signal. For a Hall element, the pulse is converted to digital and then output. For a Hall IC, it is output in the equivalent waveform. In the case of the TB6586AFG, the up-down edges of the U, V and W phase (respectively) are combined and then generated.

Example: Number of FG pulses for an 8-pole motor:

- TB6586FG: 4 pulses per revolution (4 ppr)
  - TB6586AFG: 12 pulses per revolution (12 ppr)



#### FG Signal Timing Chart

#### 8. Protecting input pin

(1) Overcurrent protection (Pin RS)

When the DC link current exceeds the internal reference voltage, this pin performs gate block protection. Overcurrent protection is restored for each carrier frequency. The pin is equipped with a filter (analog filter = 0.5  $\mu$ s (typ.)) that prevents malfunctioning due to external noise.

- (2) Position detection signal error protection When the position detection signals are either all High, Low or Open, all the output is turned OFF (all phases Low). Anything else results in a restart.
- (3) Low power voltage protection (V<sub>CC</sub> power monitor)

If the operation voltage range is exceeded when the power is being turned on or off, all the output is turned Low to prevent short circuit damage to the power element. Also, if 2.1 V or higher is input via the VSp pin, and if the motor is not rotating (Hall signal = 5 Hz or less), then normal drive is restored after a refresh operation (1.5 ms (typ.)) is performed. However, operations cannot be guaranteed during a power restoration as the circuitry will be unstable when the power is turned on.



#### (4) Output pulse width restriction

To prevent damage to the output driver (externally attached), the drive output signals (UH, VH, WH, UL, VL, WL) are restricted from being output at a pulse width of 1  $\mu$ s or less.

(5) Reset circuit

When 1.7 V (typ.) or more is input to the RESET pin, a reset will be performed with all output phases being turned off (i.e., all phases Low). Output is also turned off if 6.35 V (typ.) or more is supplied to the CW/CCW pin. However, do not use this method as the restoration obtained from it is unstable.

• RESET pin: Output off reset

All output phases are turned Low and the externally connected power element is stopped. When 1.7 V or less is input, the power is restored. During the restoration, if 2.1 V or more is not input to the Vsp pin, and if the motor is not rotating (Hall signal = 5 Hz or less), a refresh operation will be performed (1.5 ms (typ.)). Normal drive will then be restored.

During the reset, the internal counter continues to operate and the FG signal continues to be output.

• CW/CCW pin: System reset

All output phases are turned Low and the externally connected power element is stopped. Restoration takes place at an input of 6.35 V (typ.). However, operation after this kind of system reset is unstable.

TB6586FG: During a system reset, the FG signal is output in compliance with the U-phase Hall signal.

TB6586AFG: The FG signal is not output during a system reset.



### Timing Chart (CW/CCW = High, LA = GND)



\*: When CW/CCW = Low and a reverse Hall signal is input, it runs at 120° energization for a lead angle of 0° ("headwind" operation).



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### Timing Chart (CW/CCW = Low, LA = GND)



\*: When CW/CCW = Low and a reverse Hall signal is input, the motor runs at 120° energization for a lead angle of 0° ("headwind" operation)

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#### **Example Application Circuit**



Note: Utmost care is necessary in the design of the output, V<sub>CC</sub>, V<sub>M</sub>, and GND lines since the IC may be destroyed by short-circuiting between outputs, air contamination faults, or faults due to improper grounding, or by short-circuiting between contiguous pins. Add overcurrent protection such as a fuse to make the device drive normally should a current exceeding the maximum rating flow in the IC for any reason.

### Package Dimensions



Weight: 0.27 g (typ.)

#### **Notes on Contents**

#### 1. Block Diagrams

Some of the functional blocks, circuits, or constants in the block diagram may be omitted or simplified for explanatory purposes.

#### 2. Equivalent Circuits

The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.

#### 3. Timing Charts

Timing charts may be simplified for explanatory purposes.

#### 4. Application Circuits

The application circuits shown in this document are provided for reference purposes only. Thorough evaluation is required, especially at the mass production design stage.

Toshiba does not grant any license to any industrial property rights by providing these examples of application circuits.

#### 5. Test Circuits

Components in the test circuits are used only to obtain and confirm the device characteristics. These components and circuits are not guaranteed to prevent malfunction or failure from occurring in the application equipment.

#### IC Usage Considerations Notes on handling of ICs

[1] The absolute maximum ratings of a semiconductor device are a set of ratings that must not be exceeded, even for a moment. Do not exceed any of these ratings.

Exceeding the rating(s) may cause the device breakdown, damage or deterioration, and may result injury by explosion or combustion.

[2] Do not insert devices in the wrong orientation or incorrectly.

Make sure that the positive and negative terminals of power supplies are connected properly.

Otherwise, the current or power consumption may exceed the absolute maximum rating, and exceeding the rating(s) may cause the device breakdown, damage or deterioration, and may result injury by explosion or combustion.

In addition, do not use any device that is applied the current with inserting in the wrong orientation or incorrectly even just one time.

#### Points to remember on handling of ICs

Heat Radiation Design

In using an IC with large current flow such as power amp, regulator or driver, please design the device so that heat is appropriately radiated, not to exceed the specified junction temperature (T<sub>J</sub>) at any time and condition. These ICs generate heat even during normal use. An inadequate IC heat radiation design can lead to decrease in IC life, deterioration of IC characteristics or IC breakdown. In addition, please design the device taking into considerate the effect of IC heat radiation with peripheral components.

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